PATSI — A Block Diagram Compiler for TX-2
Programming Aid To System Investigation

PATSI is a TX-2 compiler for system simulation. The simulation of a system — composed of elements like filters, adders, multipliers, gates, delays and the like — can be easily programmed from the block diagram of the system. A useful feature is the ability to connect a scope to any or all of the waveforms in the system, and observe the progress of the simulation. The user may also interact with his system through the use of various knobs and switches on the TX-2 console.

PATSI has been used in the simulation of several speech-compression devices, and has considerably lessened the burden of programming them.

A typical PATSI statement (one line of typing) describes one block or element in the block diagram. It must give the element a name, or tag; it must tell what type of element the block is; and it must specify the parameters of the block, such as the input(s), gain, frequencies, etc.

TAG → ELEMENT TYPE | PARAMETERS

 $Z \rightarrow ADDER \mid X, Y, W13$ 

For example,

describes an element called Z whose output is the sum of its inputs. These inputs are the outputs of the elements called X, Y, and W13.

The tag serves three purposes. It identifies the line in the PATSI program, it serves as the name of the element, and it is the name of the output of the element.

For a given element type, i.e., COSINE GENERATOR, ADDER, DELAY, the form of the statement, i.e., the order and meaning of the parameters, is found in the "DICTIONARY," along with the limitations and restrictions of the particular block. The dictionary form of ADDER, for example, is

ADDER | in1, in2, in3,....

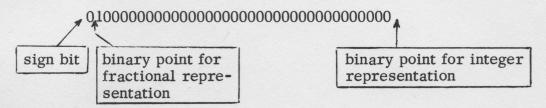
The last line of every program is "DONE."

The following points should be kept in mind in the specification of parameters as numbers:

1. A number typed will be treated as base 8 (instead of base 10) unless followed by a period. Thus, when we mean a number to be decimal system, as we usually

do, it is followed by a period.

- 2. A number typed is treated as an integer unless preceded by a period. Mixed numbers are not allowed.
- 3. A collection of bits in a register has no intrinsic numerical value until the position of the decimal point (binary point) is specified. The binary point position must be specified or assumed before a collection of bits can be called equal to a number. In rule 2, "treated as" an integer means that the binary point in the register is taken to be after the least significant bit. "Treated as" a fraction means the binary point is considered to be in front of the most significant bit. Thus, typing .5. results in the same collection of bits as typing 17179869184. (which is  $2^{34}$ ). The binary representation of each is



4. Once a number has been represented in a register as a collection of bits, the location of the binary point is lost. However, each element will deal with parameters as if the decimal point were where it is expected. Thus, GAIN expects an integer, whereas ATTENUATE expects a fraction. Where the DICTIONARY does not suggest a specific form for the parameter, the element probably can operate on the number in the register without needing to know where the binary point is. In this case, it will assume that the binary point in the parameter register is in the same position as the binary point in the waveform register. For example, CLIP | in, . 5. > . 25. is exactly equivalent to

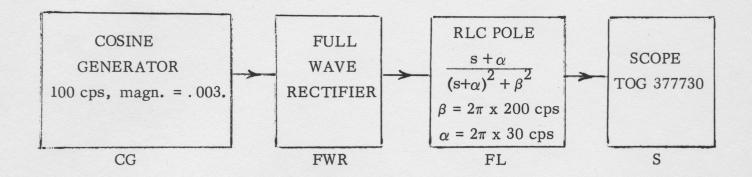
5. Register addresses are most easily specified as octal integers (as 377723).

## Lights and Meanings

When the system simulated runs into trouble such as machine overflow, or when the system does something which results in an impossible situation, such as presenting the square root taker with a negative input, certain lights <u>may</u> be lit up to warn the user. Only the lights mentioned below are used. Other troubles may not be detected.

Light Number	Meaning
1	Output has filled core region allotted to it.
2	Modulator output has clipped.
3	Gain output has clipped.
4	Divider output has clipped.
5	Tried square root of negative quantity.
6	Not used.
7	Overflow for adder or difference.
8	Not used.
9	Not used.

## Example:



Let the sampling rate be 10000/sec. The PATSI program is:

CG → COSINE GENERATOR | .01., .003.

FWR → FULL WAVE RECTIFIER | CG

 $FL \rightarrow RLC POLE \mid FWR, .003., .02.$ 

 $S \rightarrow SCOPE \mid FL, 377730$ DONE

OUDER OF COMPILED PROGRAM SAME AS ORDER OF SYMBOLIC PROGRAM

## PATSI Dictionary

- I. Elements the number in parenthesis is the page on which a description of the element may be found.
  - (9) ADDER
  - (9) ATTENUATE
  - (10) BAND PASS FILTER
  - (11) CLIP
  - (12) CONTROLLED OSCILLATOR (square wave)
  - (10) CONVOLVE
    - (7) COSINE GENERATOR
    - (9) DELAY
    - (9) DIFFERENCE
  - (11) DIVIDER
  - (10) EXPOSINE (h(t) =  $e^{-\alpha t} \sin \beta t$ ; H(S) =  $\frac{\beta}{(s+\alpha)^2 + \beta^2}$ )
  - (11) FULL WAVE RECTIFIER
  - (9) GAIN
  - (11) GATE
  - (11) GREATER OF
  - (11) HALF WAVE RECTIFIER
  - (7) INPUT (from core memory)
  - (9) INVERTER
  - (11) LIMITER
  - (11) MODULATOR (multiplier)
    - (7) NOISE GENERATOR
  - (12) ONE SHOT (monostable multivibrator)
    - (8) OUTPUT (to core memory)
  - (11) PEAK DETECTOR
  - (7) PERIODIC INPUT (from core memory)
  - (10) RC POLE (h(t) =  $\alpha e^{-\alpha t}$ ; H(S) =  $\frac{\alpha}{s + \alpha}$ )
  - (10) RLC POLE (h(t) =  $e^{-\alpha t} \cos \beta t$ ; H(S) =  $\frac{s + \alpha}{(s + \alpha)^2 + \beta^2}$ )
  - (10) RLC ZERO (H(S) =  $\frac{(s+\alpha)^2 + \beta^2}{\beta}$ )
  - (12) SAMPLE AND HOLD

- (8) SCOPE
- (9) SCOPE SYNC
- (12) SQUARE ROOT
- (12) SWITCH
- (7) VARIABLE RATE PULSER (hand controlled)
- (8) XYSCOPE
- (11) ZERO CROSSING PULSER

II. Control Statements - these are not elements, although they may relate to them. The parenthsis again contain page numbers.

for moveable poles and zeros

- (13) BRANCH UNLESS
- (13) BREAK
- (7) CHANGE COSINE FREQUENCY
- (10) CHANGE EXPOSINE
- (10) CHANGE RC POLE
- (10) CHANGE RLC POLE
- (10) CHANGE RLC ZERO
- (13) MULTIPLY T BY
- (13) RETURN

33 DEFINE

33 EMD

INPUT  $| \alpha \rightarrow \beta$ 

use 5-wemony for Input/output

The output of this block at successive sampling times is the contents of successive registers of core memory.

 $\alpha$  = first  $\beta$  Register of the area of core memory containing the desired input waveform.

If the data is to be configured, the number of the configuration is superscripted after  $\beta,~e.\,g.$  , INPUT  $|~1\to100000^{12}$ 

When  $\alpha \to \beta$  is used up (after  $(\beta - \alpha)$  samples), we go to MKIV.

PERIODIC INPUT  $| \alpha \rightarrow \beta$ 

This differs from the above only in that when  $\alpha \to \beta$  is used up, we return to  $\alpha$ . Thus the waveform in  $\alpha \to \beta$  is repeated endlessly.

COSINE GENERATOR | F, M

F = cosine frequency as a fraction of the sampling frequency.

M = amplitude. It should be large, or the difference equation used will not be effective.

F should be greater than .00003.

The output of this block is a sample of  $\cos 2\pi F/F_s t$ . The corresponding sine wave ( $\sin 2\pi F/F_s t$ ) is found at X + 2 where X is the tag of the cosine generator.

NOISE GENERATOR

No parameters. The output is a pseudorandom uniformly distributed noise sequence.

VARIABLE RATE PULSER | M

The output is normally  $0^{\circ}$ . It is replaced with a one sample pulse of amplitude M periodically. The repetition rate is given by the contents of the left half of the KNOB.

f of the KNOB.

pulse frequency =  $\frac{\text{KNOB}_{3,4}}{4\ 000\ 000}$  x sampling frequency.

CHANGE COSINE FREQUENCY | X,  $\omega$ 

X is the tag of the cosine generator affected.  $\omega$  is the tag of an element whose

output is to control the frequency of X. The output of  $\omega$  is taken as a fraction of the sampling frequency.

This operation is slow and should be done only once per several sampling times if possible.

OUTPUT | in, 
$$\alpha \rightarrow \beta$$

This element allows the waveform at "in" to be saved in core memory.  $\alpha$  and  $\beta$  are as for INPUT. Configuration is allowed.  $\alpha$  and  $\beta$  must be in S memory ( $\alpha < \beta < 200000$ ). When  $\alpha \to \beta$  is used up, new samples are ignored, and push button 1 lights up.

in = tag of element whose output is fed to the scope.

CT = control toggle, used as follows:

4.10 
$$\begin{cases} 0 & \text{show waveform} \\ 1 & \text{do not show waveform} \end{cases}$$

Q3 - sweep frequency. The number of samples corresponding to the scope face is given by  $\frac{100000}{\text{O3}}8$  .

Q2, Q4 are amplitude controls. Gain of scope =  $.(Q2) \times 2^{(Q4)}$ .

Q1 = vertical position control.

If several scopes are used, all the sweep rates are added together. Good practice is to set all but one to zero so that sweep rate control is by only one toggle register. (See scope sync.)

x = tag of horizontal input.

y = tag of vertical input.

Cx = Control toggle for horizontal input, <math>Cy = control toggle for vertical input, as follows:

4. 10 as for SCOPE

Q2, Q4 as for SCOPE

Q1 = horizontal position control for Cx, vertical position control for Cy.

Q3 - not used.

SCOPE SYNC | X > CT

The output of X, and control toggle CT are used to control the scope as follows:

4. 10 of CT = 
$$\begin{cases} 0 & \text{no sync} \\ 1 & \text{see below} \end{cases}$$

If the scope trace has not reached the end of the screen nothing happens. If the scope trace has reached the end of the screen, the scope is turned off until the output of X is greater than the number in CT. When this happens, the scope is turned on again, and the trace reset to the left side of the scope face.

DELAY | in, N

The output of this element is the same as the output of "in" but delayed by N sampling intervals. N is a positive integer. DELAY uses some memory from 215777 down, as necessary.

GAIN | in, M

The output is the same as the output of "in" but with a gain of M. M is an integer, positive, negative, or zero. If the product is too big,  $\pm$  37777777777 is used, and push button 3 is lit.

ATTENUATE | in, k

Like GAIN, but k is a fraction, positive, negative, or zero.

The combination of GAIN and ATTENUATE can give any fixed multiplier.

INVERTER | in

A gain of -1.

ADDER | in1, in2, in3, ....

The output is the sum of the input waveforms. Overflow is detected and indicated by lighting push button #7, but not corrected. The number of inputs is limited to 12.

DIFFERENCE | in1, in2

The output is (in1) - (in2). Overflow is as for ADDER.

RC POLE | in, F

Sampled data equivalent of H(S) =  $\frac{\alpha}{S+\alpha}$  . Here F is  $\alpha$  divided by the sampling frequency.

RLC POLE | in, F, G

Sampled data equivalent of H(S) = 
$$\frac{S + \alpha}{(S+\alpha)^2 + \beta^2}$$
. Here

F is  $\alpha$  divided by the sampling frequency. G is  $\beta$  divided by the sampling frequency.

EXPOSINE | in, F, G

Sampled data equivalent of H(S) =  $\frac{\beta}{(S+\alpha)^2 + \beta^2}$   $\alpha$  and  $\beta$  related to F, G as in RLC POLE.

RLC ZERO | in, F, G

Sampled data equivalent of H(S) =  $\frac{(S+\alpha)^2 + \beta^2}{\beta}$   $\alpha, \beta$  related to F, G as in RLC POLE.

CONVOLVE | in, 
$$\alpha \rightarrow \beta$$

 $\alpha$  is the first and  $\beta$  is the last register of an area of core memory containing the function to be convolved with the waveform at "in." This is brute force simulation of filters, and is very slow. Some memory is used from 215777 down as needed.

A cascade of up to 6 <u>RLC Poles.</u> In addition to the theoretical delay, there is a delay of one sampling interval for each RLC POLE used after the first.

CHANGE RLC POLE 
$$\mid X, \alpha, \beta$$
  
CHANGE EXPOSINE  $\mid X, \alpha, \beta$   
CHANGE RLC ZERO  $\mid X, \alpha, \beta$   
CHANGE RC POLE  $\mid X, \alpha$ 

These permit moveable poles and zeros.  $\alpha$  and  $\beta$  are tags of control elements, and X is the tag of the element whose poles or zeros are to be moved. The output of  $\alpha$  corresponds to F and the output of  $\beta$  corresponds to G.

These are slow, and should be used in conjunction with "MULTIPLY T BY  $\mid$  N" when possible.

GATE | in, C in

"in" is the waveform to be gated. Cin is the control waveform.

output = 
$$\begin{cases} 0 & \text{Cin } \le 0 \\ \text{(in)} & \text{Cin } > 0 \end{cases}$$

GREATER OF | in1, in2

The output is the greater of the two inputs.

HALF WAVE RECTIFIER | in FULL WAVE RECTIFIER | in

These are self-explanatory. The HALF WAVE output is  $0^-$  for negative input.

ZERO CROSSING PULSER | in, M

The output is  $0^-$  except following a zero crossing when it is M. The direction of the zero crossing is not noted.

PEAK DETECTOR | in, M

The output is normally  $0^+$ . Following a positive peak ( $\bigwedge$ ), the output is a pulse of height M. Following a negative peak ( $\bigvee$ ), the output is a pulse of height -M.

LIMITER | in, M  
output = 
$$\begin{cases} M, & \text{in } \ge 0^+ \\ -M, & \text{in } \le 0^- \end{cases}$$

CLIP | in, T > B

output = 
$$\begin{cases}
T, & \text{in } \ge T \\
\text{in, } & \text{T } \ge \text{in } \ge B \\
B, & \text{B } \ge \text{in}
\end{cases}$$

T and B are fixed levels, with T > B.

MODULATOR | in1 x in2 DIVIDER | in1/in2

The output of MODULATOR is the product of the inputs, scaled appropriately. The product of fractions is scaled 17. places to the left. The product of integers are scaled 19. places to the right. DIVIDER is the inverse, so a ratio of fractions is

scaled 17. places to the right, and a ratio of integers is scaled 19. places to the left. Outputs which are too large are replaced by  $\pm$  377777777777 and push button #2 for the modulator or #4 for the divider are lit.

SQUARE ROOT | in

The input is treated as a fraction. Thus  $\sqrt{\text{in}} \ge \text{in}$ . If (in) is negative, the square root of the magnitude is taken, and push button #5 is lit.

SWITCH | Cin, POSin, NEGin

The output is taken from 
$$\begin{cases} POSin & Cin \ge 0^{+} \\ NEGin & Cin \ge 0^{-} \end{cases}$$

ONE SHOT | Sin, M, ONTIME

This has two states. In the off state, the output and next state are:

$$\begin{cases} 0^{-} \text{ and off if Sin} \leq 0 \\ M \text{ and on if Sin} > 0 \end{cases}$$

ONTIME = 21??

In the on state, the output and next state are:

M and off if the ON state has lasted "ONTIME" sampling intervals.

M and on if the ON state has not lasted "ONTIME" sampling intervals.

ONTIME is an integer greater than 0.

SAMPLE AND HOLD | in, Cin

Output = 
$$\begin{cases} (in), & Cin > 0 \\ last output, & Cin \le 0 \end{cases}$$

CONTROLLED OSCILLATOR | in, M

The output is a square wave of amplitude M and frequency given by the waveform at "in," according to the formula

$$f = \frac{.(in)}{2}$$
 x sampling frequency

BREAK : RETURN

These are used together. A conventional program placed in between will be executed once each time around the loop.

## MULTIPLY T BY N

The program or elements following this line are changed or performed only once every N times around the loop. Thus, they seem to have a sampling rate of F, or a sampling interval of NT. N is a positive integer.

These may be nested so that program parts following two of these statements are looked at every MxN times around the loop.

BRANCH UNLESS | 
$$F = \alpha, \rightarrow X$$

The simulation will interrupt the normal sequence of control and proceed to line X, except if the output of F is equal to  $\alpha$ .

For example, to show a dashed line representation of a function on the scope:

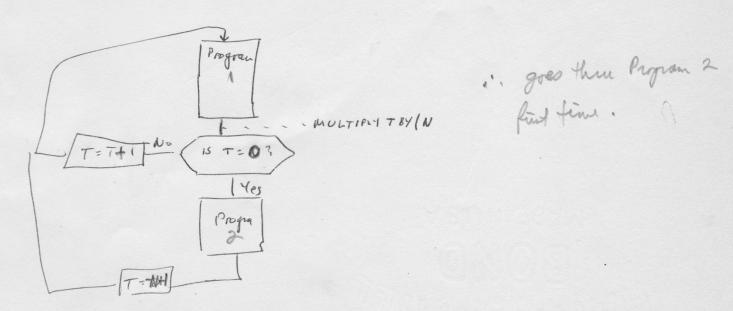
$$MX \rightarrow COSINE GENERATOR \mid F, .1$$

$$BRANCH UNLESS \mid X = 100, \rightarrow X$$

$$SCOPE \mid Y, CT$$

$$X \rightarrow LIMITER \mid MX, 100$$

PS - This assumes that "sweep rate" is given by some other scope which is looked at every time around the loop.



FFREP STARTΔ→SSΔγ

START → STE BACK A \*

\*\*SAVE MARK 4 RETURN ADDRESS

RFD 60 ABEGIN

FFINS SKIPNSEQA

REX 47 INPUTAY \*\*INITIALIZE SEQ 47

IOS4730000 \*\*CONNECT MISC INPUTS

FFINS DSALAY

INPUTAy→ STE P \*\*SEQ 47 START POINT. SAVE E.

hTSD XXAY

3 STE #+1

110S<sub>75</sub>30000 \*\*SET LIGHTS = INPUT BUTTON

SKZ1.9XXAY \*\*BUTTON 9 MEANS QUIT

JMP QUITAY

SKZ1.8XXAY \*\*BUTTON 8 MEANS PRINT, IF MACRO USED

DUMPAy→ JMP #+1 \*\*RESET BY DUMPA MACRO IF USED

LDE P

JPD INPUTAY \*\*RETURN TO PATSI

QUITAY→ IOS 60 20000 \*\*DISCONNECT SCOPE AND LOWER FLAG

105 60 40000

QUIT1△→ RFD<sub>70</sub>#+1

REX 47 INPUTAY \*\* RESET SEQ 47 COUNTER

BACK△→ JPQ ## \*\*RETURN TO MARK 4

FIFEND

33 REP CONTINUE +4 &Br - RETURN

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FRE OUIX PATST CITTING 6/64

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DORIG!

200000 000000 200204

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021 001660 600051

023 43,0601 600337

SKZ 4 . 10 \*COND2 LDA \*CONDO JMP #+13 FULLSCOPAY 1 JMOL 30ADD 35LDA 36ADD SAB SCOPA

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in, Escopo3, 70h

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132

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